## 5. The centroid is given as

Centroid = 
$$\frac{\text{Sum of poles - Sum of zeros}}{P - Z} = \frac{(0 - 1 - 5) - (-1.5)}{3 - 1} = -2.25$$

### 6. Break-away points:

The C.E. of the system is given as

$$1 + G(s)H(s) = 0$$

$$1 + \frac{K(s+1.5)}{s(s+1)(s+5)} = 0$$

$$K = \frac{s(s^2 + 6s + 5)}{s+1.5}$$

$$\frac{dK}{ds} = 0 \Rightarrow s^3 + 5.25s^2 + 9s + 3.75 = 0$$

$$s = 0.6, s = 2.3 + j0.89, s = -2.3 - j0.89.$$

The valid break-away point is  $B_1 = -0.6$ 

### 7. Angle of departure:

Since there is no complex pole and zero, we need not find the angle of departure or arrival.

### 8. Intersection points of the root locus branches with imaginary axis:

The C.E. of the system is given as

$$1 + G(s)H(s) = 0$$

$$1 + \frac{K(s+1.5)}{s(s+1)(s+5)} = 0$$

$$\Rightarrow \qquad s(s^2 + 6s + 5) + K(s+1.5) = 0$$

$$\Rightarrow \qquad s^3 + 6s^2 + (5+K)s + 1.5K = 0$$

$$\begin{vmatrix} s^3 \\ 5^2 \\ 6 \\ 1.5K \\ 0 \\ 0 \end{vmatrix}$$

$$\begin{vmatrix} 1 \\ 5+K \\ 0 \\ 0 \\ 1.5K \\ 0 \\ 0 \end{vmatrix}$$

For a stable system, 1.5K > 0, (30 + 4.5K)/6 > 0

$$\Rightarrow$$
  $K > 0, K < -6.67$ 

Since the value of K is negative, there is no crossing point on the imaginary axis. In other words for any positive value of K, the root locus will not cross the imaginary axis. The complete root locus plot is as shown in Figure (m).

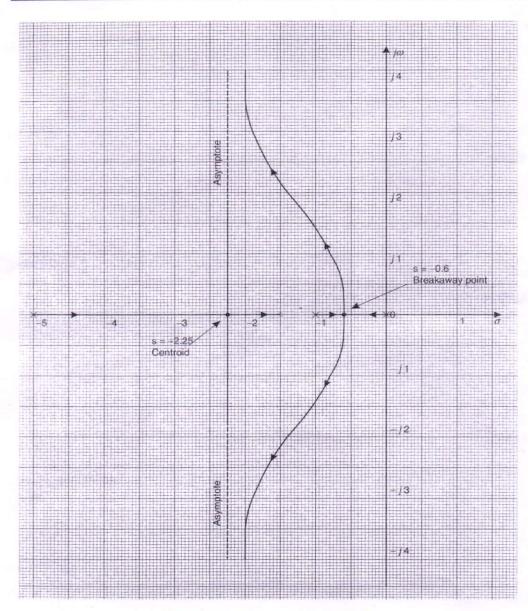


Figure (m) Root locus plot for ASP-7

**ASP-8:** Sketch the root locus plot of a unity feedback system whose OLTF is  $G(s) = \frac{s}{(s^2 + 4)(s + 2)}$ .

Solution:

Given that

$$G(s) = \frac{s}{(s^2+4)(s+2)}.$$

- 1. Root locus is symmetrical about real axis.
- 2. The root locus plot starts from a pole at K=0 and terminates at an open-loop zero, that is, at  $K=\infty$ .
- 3. The poles are given as

$$s = -2$$
,  $s^2 + 4 = 0$ 

$$P_1 = -2$$
,  $P_2 = j2$ ,  $P_3 = -j2$ ,

The zeros are given as

$$s = 0$$

Number of poles = P = 3

Number of zeros = Z = 1

Number of root locus branches = N = P = 3

Number of asymptotic lines = n = P - Z = 2

4. The angle of asymptotic lines with negative real axis

$$\phi = \frac{(2q+1)180}{P-Z}$$
, where  $q = 0, 1$   
= 180/2, 3 × 180/2  
= 90, 270.

5. The centroid is given as

Centroid = 
$$\frac{\text{Sum of poles - Sum of zeros}}{P - Z} = \frac{(0 - 2 - j2 + j2) - (0)}{3 - 1} = -1.$$

The complete root locus is as shown in the below figure.

**ASP-9:** Sketch the root locus plot of the system whose OLTF is given as  $G(s)H(s) = \frac{K}{s(s+4)(s^2+8s+32)}$ 

Hence, find the value of K so that the system has a damping factor 0.707.

#### Solution:

Given that

$$G(s)H(s) = \frac{K}{s(s+4)(s^2+8s+32)}$$

- 1. Root locus is symmetrical about real axis.
- 2. The root locus plot starts from a pole at K=0 and terminates at an open-loop zero, that is, at  $K=\infty$ .
- 3. The poles are given as

$$s = 0$$
,  $s = -4$ ,  $s^2 + 8s + 32 = 0$ 

$$P_1 = 0$$
,  $P_2 = -4$ ,  $P_3 = -4 + j4$ ,  $P_4 = -4 - j4$ , Number of poles  $= P = 4$ 

Number of zeros = Z = 0

Number of root locus branches = N = P = 4

Number of asymptotic lines = n = P - Z = 4

4. The angle of asymptotic lines with negative real axis

$$\phi = \frac{(2q+1)180}{P-Z}, \text{ where } q = 0, 1, 2, 3$$
$$= 180/4, 3 \times 180/4, 5 \times 180/4, 7 \times 180/4$$
$$= 45, 135, 225, 315.$$

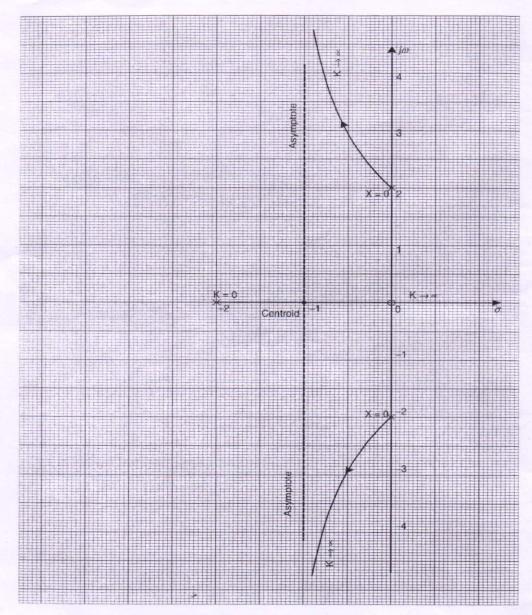


Figure (n) Root locus for ASP-9

# 5. The centroid is given as

Centroid = 
$$\frac{\text{Sum of poles} - \text{Sum of zeros}}{P - Z} = \frac{(0 - 4 - 4 + j4 - 4 - j4) - (0)}{4 - 0} = -3.$$

The complete root locus is as shown in Figure (n).

### 6. Break-away points:

The C.E. of the system is given as

$$1 + G(s)H(s) = 0$$

$$1 + \frac{K}{s(s+4)(s^2+8s+32)} = 0$$

$$K = -s(s+4)(s^2+8s+32)$$

$$K = -(s^4+12s^3+64s^2+128s)$$

$$\frac{dK}{ds} = 0 \Rightarrow 4s^3+6s^2+128s+128 = 0$$

$$s^3+9s^2+32s+32 = 0.$$

After solving the above-mentioned equation, the valid break-away point is  $B_1 = -1.57$ .

# 7. Intersection points of the root locus branches with imaginary axis:

The C.E. of the system is given as

$$1 + G(s)H(s) = 0$$

$$s^{4} + 12s^{3} + 64s^{2} + 128s + K = 0$$

$$s^{4} \begin{vmatrix} 1 & 64 & K \\ s^{3} & 12 & 128 & 0 \\ 53.33 & K & 0 \\ s^{1} & 128 - 0.225K & 0 & 0 \\ s^{0} & K & 0 & 0 \end{vmatrix}$$

For a stable system, K > 0, 128 - 0.225K > 0

$$\Rightarrow$$
 0 < K < 569.

For a stable system, the maximum value of K is 569. For K > 569, the roots lie on the RHS of the s-plane, and hence, K = 569 is the value for marginal stability.

The auxiliary equation is

$$53.33s^2 + K = 0$$

$$53.33s^2 + 569 = 0$$

$$s = \pm j \ 3.266.$$

### 8. Angle of departure:

$$\begin{aligned} \phi_{p3-(-4+j4)} &= 180 - (\phi_{p1} + \phi_{p2} + \phi_{p4}) \\ &= 180 - (135 + 90 + 90) = -135^{\circ} \\ \phi_{p4-(-4-j4)} &= +135^{\circ} \end{aligned}$$

The complete root locus plot is as shown in Figure (o).

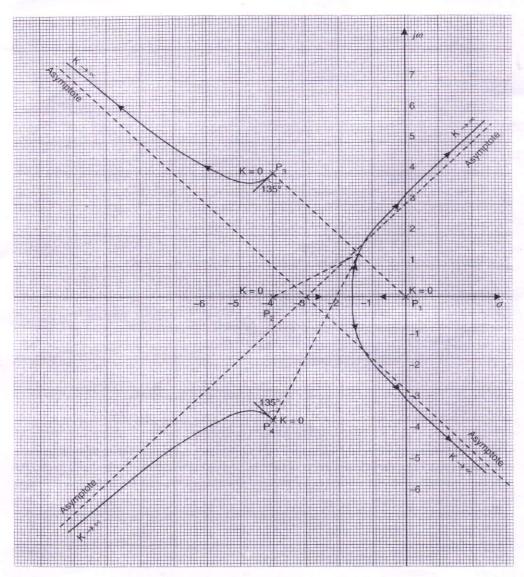


Figure (o) Root locus plot for ASP-9

Since  $\xi = 0.707 \Rightarrow \cos \theta = \xi = 0.707 \Rightarrow \theta = 45^{\circ}$ From the origin, a line at an angle  $\theta = 45^{\circ}$  is drawn, which intersects the root locus plot at s = -1.35 + 1.00j1.35. As this point lies on the root locus, the following condition is satisfied.

$$|G(s)H(s)| = 1$$

$$\left| \frac{K}{s(s+4)(s^2+8s+32)} \right| = 1$$

$$\left| \frac{K}{(-1.35 + j1.35)(-1.35 + j1.35 + 4)[(-1.35 + j1.35)^{2} + 8(-1.35 + j1.35) + 32]} \right| = 1$$

$$\frac{k}{130} = 1 \Rightarrow K = 130.$$

**ASP-10:** The OLTF of a unity feedback system is  $G(s) = \frac{K(s+a)}{s(s+b)}$ .

1. Prove that break-away and break-in points will exist only when |a| > |b|.

2. Prove that the complex points on the root locus form a circle with centre (-a, 0) and radius  $\sqrt{a^2 - ab}$ .

#### Solution:

Given that

$$G(s)H(s) = \frac{K(s+a)}{s(s+b)}.$$

There are two open-loop poles, and hence, the number of branches are 2 and start at  $P_1 = 0$  and  $P_2 = -b$ , respectively. One of the branches terminates at infinity and other terminates at zero  $Z_1 = -a$ .

1. The C.E. is given as

$$1 + G(s)H(s) = 0$$

$$1 + \frac{K(s+a)}{s(s+b)} = 0$$

$$K = -\frac{s(s+b)}{s+a}$$

$$\frac{dK}{ds} = 0 \Rightarrow s^2 + 2as + ab = 0$$

$$s = -a \pm \sqrt{a^2 - ab}.$$
(2)

Substituting the value of s in (1), we will get

$$K = -\frac{\left[-a + \sqrt{a(a-b)}\right]^{2} + b\left[-a + \sqrt{a(a-b)}\right]}{-a + \sqrt{a(a-b)} + a}$$

$$= \frac{a^{2} + a(a+b) - 2a\sqrt{a(a-b)} + \left[-ab + \sqrt{a(a-b)}\right]}{\sqrt{a(a-b)}}$$

$$= -\left[2\sqrt{a(a-b)} + b - 2a\right].$$

From the above-mentioned equation of K, it is clear that for all values of a and b such that |a| > |b|, K is positive. Therefore, the break-away or break-in points will exists only when |a| > |b|.

1. If |a| > |b|, then break-away and break-in points exist. Let us choose a test point in the complex plane s = x + jy that lies on the root locus. This point should satisfy angle criterion so that the point is a part of the root locus.

$$G(s)H(s)_{s-x+jy} = \frac{K(x+jy+a)}{(x+jy)(x+jy+b)} \ .$$

The phase angle is given as

$$\angle G(s)H(s)_{s=-x+jy} = \tan^{-1}\frac{y}{x+a} - \tan^{-1}\frac{y}{x} - \tan^{-1}\frac{y}{x+b} = -180^{\circ}$$

$$180 + \tan^{-1} \frac{y}{x+a} = \tan^{-1} \frac{y}{x} + \tan^{-1} \frac{y}{x+b}$$

By taking 'tan' on both sides, we have

$$\frac{\frac{y}{x} + \frac{y}{x+b}}{1 - \frac{y}{x} \times \frac{y}{x+b}} = \frac{\tan 180 + \frac{y}{x+a}}{1 - \tan 180 \times \frac{y}{x+a}}$$

After simplifying the above-mentioned equation, we have

$$x^2 + 2xa + y^2 = -ab$$

By adding both sides  $a^2$ , we get  $(x+a)^2 + y^2 = (a^2 - ab)$ 

$$(x+a)^2 + y^2 = (a^2 - ab)$$

$$(x+a)^2 + (y+0)^2 = \left[\sqrt{a^2 - ab}\right]^2$$
.

This is equation of a circle with centre (-a, 0) and radius  $\sqrt{a^2 - ab}$ . Hence, it is proved.

# **REVIEW QUESTIONS**

- 1. What do you understand by root locus plot?
- 2. Explain the constructional procedure of root locus plot.
- 3. Write a short note on angle and magnitude condition of root locus.
- 4. Explain the terms:
  - (i) asymptotes
  - (iii) break-away points

- (ii) centroid
- (iv) angle of departure
- 5. Define the term 'root locus' and state the rule for finding out the root locus on the real axis.
- 6. Explain the method of calculating the break-away points.
- 7. What is a root locus plot? Explain with a suitable example.
- 8. What are the features of root locus plot?
- 9. Show that the break-away and break-in points, if any, on the real axis for the root locus for  $G(s)H(s) = \frac{KN(s)}{D(s)}$ , where N(s) and D(s) are rational polynomials in S, and that they can be obtained by solving the equation  $\frac{dK}{ds} = 0$ .
- 10. Explain the stability of the system from the root locus plot in the following situations with suitable examples:
  - (a) addition of open-loop poles.
  - (b) addition of open-loop zeros.
- What is break-away and break-in point? How are they determined?
- 12. What is centroid? How the centroid is calculated?
- 13. Open-loop t.f. of a unity feedback system is  $G(s) = \frac{K(s+a)}{s(s+b)}$ (a) Prove that break-away and break-in points will exist only when |a| > |b|

  - (b) Prove that the complex points on the root locus form a circle with centre (-a, 0) and radius  $\sqrt{a^2 - ab}$

## **EXERCISE PROBLEMS**

- 1. Check whether the point s = -3 + j5 lies on the root locus of a system having  $G(s)H(s) = \frac{K}{(s+1)(s+5)}$ . Determine the corresponding value of K.
- 2. Sketch the root locus for the system having the OLTF,  $G(s)H(s) = \frac{K(s+4)(s+5)}{(s+1)(s+3)}$ .
- 3. A unity feedback system has an OLTF,  $G(s) = \frac{K}{s^2(s+2)}$ .
  - 1. By sketching a root locus plot, show that the system is unstable for all values of K.
  - 2. Add a zero to  $s = -a(0 \le a \le 2)$  and show that the addition of zero stabilises the system.
- **4.** A feedback control system has an OLTF  $G(s)H(s) = \frac{K}{s(s+3)(s^2+2s+2)}$ . Find the root locus as K is varying from 0 to  $\infty$ .
- 5. Sketch the root locus for the system whose OLTF is given as  $G(s)H(s) = \frac{K(s+4)(s+5)}{(s+3)(s+1)}$
- 6. Sketch the root locus plot for all the values of K from 0 to  $\infty$  for a negative feedback system whose OLTF is given as  $G(s)H(s) = \frac{K(s+1)}{s^2(s+9)}$ ?
- 7. Sketch the root locus plot for all the values of K from 0 to  $\infty$  for a negative feedback system whose OLTF  $G(s)H(s) = \frac{K(s^2 4s + 8)}{(s + 1)(s 0.5)}$ ?
- 8. Using the root locus method, find gain (K) for the system having OLTF as  $G(s)H(s) = \frac{K(s+2)(s+4)}{(s+1)(s+3)(s+5)}$  when  $\xi = 0.246$ .
- 9. Consider a negative feedback system characterised by  $G(s)H(s) = \frac{K(s+1)}{s(s+2)(s+3)(s+4)}$ . Sketch the root locus plot for all values of K ranging from 0 to  $\infty$ . Further, find gain K for which the CLTF will have a pole on the real axis at -0.5.
- 10. Sketch the root locus plot for a closed-loop system having an OLTF  $G(s)H(s) = \frac{K(s+1)(s+2)}{s(s-1)}$  for all values of K ranging from 0 to  $\infty$ . Find the range of values of K for closed-loop stability.

## **ANSWERS**

- 1. Since  $\phi = -180^{\circ}$ , the point -3 + j5 lies on the root locus, K = 29.
- 8. K = 33
- **9.** For closed-loop stability, 0 < K < 141.
- **10.** For closed-loop stability,  $0.33 < K < \infty$ .

#### **OBJECTIVE TYPE QUESTIONS**

- 1. Intersection of root locus branches with imaginary axis can be determined by
  - (a) Nyquist criterion

(b) R-H criterion

(c) Polar plot

- (d) Bode plot
- 2. The break-away point is determined by solving the equation
  - (a)  $\frac{dK}{ds} = 0$

(b) 1 + G(s)H(s) = 0

(c)  $\int \frac{dK}{ds} = 0$ 

(d) G(s) H(s) = 0

3.	The angle of asymptotic lines	s making with negat	ive rea	l axis is given as	
	(a) $\frac{2k\pi}{P-Z}$ (b)	$\frac{(2k+1)\pi}{P-Z}$	(0)	$(k+1)\pi$	(d) $\frac{(2k+1)\pi}{P-Z} \cdot \frac{\pi}{2}$
					1 4
4.	When the root locus separates	s at a point between	two o	pen-loop poles, the	point is called
	(a) critical point (b)	crossover point	(c)	centroid	(d) break-away point
5.	If the root locus branches cros	ss the imaginary axis	s, the s	system becomes	
	(a) marginally stable		(b)	conditionally stabl	e
	(c) unstable		(d)	None	
6.	The intersection of the root lo	cus branch with the	imagin	nary axis can determ	ined by using the equatio
	(a) $G(s) H(s) = 0$		(b)	1 - G(s) H(s) = 0	
	(c) $1 + G(s) H(s) = 0$		(4)	$\frac{dK}{ds} = 0$	
	(c) $1 + O(3) 11(3) = 0$		(u)	ds	
7.	The asymptotes to root locus	branches intersect o	n the	real axis at a point g	iven as
	$\sum$ Poles $-\sum$ Zeros				
	(a) $\frac{\sum Poles - \sum Zeros}{(P - Z)}$		(b)	$\sum Poles + \sum Zero$	
				(I-Z)	
	(c) $\frac{\sum \text{Poles} - \sum \text{Zeros}}{(P+Z)}$		(1)	NI	
	(P+Z)		(D)	None	
8.	In loop t.f., zeros are the sam	e as			
	(a) zeros of $G(s)H(s)$		(b)	zeros of $1 + G(s) H$	H(s)
	(c) poles of $G(s)H(s)$			poles of $1 + G(s)H$	
9.	The value of K at which the	root locus crosses the	e imag	inary axis makes the	e system
	(a) stable		(b)	under-damped	
	(c) marginally stable		(d)	unstable	
10.	The starting points of the roo	t locus are			
	(a) closed-loop poles		(b)	closed-loop zeros	
	(c) open-loop poles		(d)	open-loop zeros	
11.	The ending points of the root	locus are			1
	(a) closed-loop poles		(b)	closed-loop zeros	
	(c) open-loop poles		(d)	open-loop zeros	
12	The number of root loci bran	ches for the OITE	C(A)H	K(s+5)(	(s+6)
12.	The number of root loci bran	ches for the OLIF	3(3)11	$s(s) = \frac{1}{s(s+1)(s+3)(s+3)}$	$(s^2 + 5s + 6)$ 1s
	(a) 2 (b)	3	(c)	4	(d) 5
13.	If the root locus branches do		400		
	(a) conditionally stable	0		inherently stable	
	(c) marginally stable			unstable	
16	A : C II I	OLTE COMMO	(-)	K	
14.	A unity feedback system has a	an OLIF $G(s)H(s)$ :	$=\frac{1}{s(s^2)}$	+4s+13). The cen	troid of the asymptotes of
	and root roots prot nes at				
		-4/3			(d) -2
15.	The OLTF of an unity feedba	ck is given as $G(s) =$		5 The	e number of asymptotes of
	The root locus plot that tend t	a intinity is awen as			
	(a) 1 (b)	2	(c)	3	(d) 4
16	(a) 1 (b) The OLTF of an unity feedba root locus plot is given as	ck system is given as	G(A)	K	ha brook away noine of the
10.	root locus plot is given as	ck system is given as	(1)	s(s+1)(s+3).	ne oreak-away point of th
		-0.523		-0.7	
	(0)	0.723	(C)	-0./	(d) None

		K	
17.	A unity feedback system has an OLTF $G(s) = \frac{1}{s(s^2 - s)}$	+ 45	+13). The angle of asymptotes are given as
	(a) 450, 1,350, 2,250	(b)	600, 1,800, 3,000
18.	(c) 900, 1,800, 2,700 The OLTF of a feedback system is $\frac{K}{s(s+6)(s^2+4s)}$	+13	The four branches of root locus originate at
19.	(c) $0, -6, -2 + j3, -2 - j3$ When the open-loop poles are added to the system	(b) (d) (b) (d) (b)	None  settling time increases All settling time increases
21.	The number of root loci of a unity feedback system is  (a) $m$ (b) $n$		wing $m$ number of poles and $n$ number of zeros $m-n$ (d) $n-m$
22.	The root locus plot of the system having the loop break-away points		
	(a) 0 (b) 1	(c)	2 (d) 3
23.	If the C.E. of the system is $s^3 + 2s^2 + (K+1)s + 3K$	=0	, then the centroid is
24.	(a) 0.5 (b) $-0.5$ The OLTF of a feedback system is the $s = -1 + j$	(c) ang	0.25 (d) $-0.25$ gle of departure of the root locus at
25.	(a) 90 (b) -90 Root locus technique is applicable to (a) single-loop system		180 (d) -180 multiple loop system
AN	(c) both		one

1.	Ь	2.	a	3.	Ь	4.	d	5.	C	6.	c	7.	a	8.	a	9.	С	10.	C
11.	d	12.	d	13.	Ь	14.	Ь	15.	d	16.	a	17.	Ь	18.	d	19.	C	20.	d
21.	C	22.	d	23.	a	24.	d	25.	C										